

Linear Systems I

Lecture 10

Solmaz S. Kia

Mechanical and Aerospace Engineering Dept.
University of California Irvine
solmaz@uci.edu

Reading: Ch 5.1-5.3, Ch 5.5, Ch 6.1, Example 6.3 of Ref [1] (skip over the parts that cover discrete-time systems).

Note: These slides only cover part of the discussions in the class. For further details, consult your in-class notes.

- Stability
 - Internal stability of LTI systems (Lyapunov method)
 - Bounded-Input-Bounded-Output (BIBO) stability
- Controllable and Reachable subspaces

$$\begin{cases} \dot{x} = A(t)x + B(t)u, \\ y = C(t)x + D(t)u, \end{cases} \quad x(t_0) = 0_n$$

Stability addresses what happens to our solutions

- do they remain bounded
- will they get progressively smaller
- they diverge to infinity

Response is due to : $\underbrace{\text{response due to } x_0}_{\text{internal stability}} + \underbrace{\text{response due to } u}_{\text{Input-output stability}}$

Def(Bounded-input-bounded-output (BIBO) stability): A system is said to be BIBO stable if every bounded input excites a bounded output (for BIBO stability evaluation we analyze the zero-state response of the system).

An input $u(t)$ is **said to be bounded** if $u(t)$ does not grow to positive or negative infinity, or equivalently, \exists a constant u_m s.t.

$$|u(t)| \leq u_m < \infty, \quad \forall t \geq 0.$$

$$\begin{cases} \dot{x} = Ax + Bu, \\ y = Cx + Du, \end{cases} \quad x(t_0) = 0_n \quad (*)$$

$$y_f(t) = y_{zs}(t) = \int_0^t g(t-\tau)u(\tau)d\tau = \int_0^t g(\tau)u(t-\tau)d\tau$$

Theorem

A SISO system (*) is BIBO if and only if $g(t)$ is absolutely integrable in $[0, \infty)$ or

$$\int_0^{\infty} |g(t)|dt \leq M < \infty$$

$$\begin{cases} \dot{x} = Ax + Bu, \\ y = Cx + Du, \end{cases} \quad x(t_0) = 0_n \quad (\star)$$

$$y_f(t) = y_{zs}(t) = \int_0^t Ce^{A(t-\tau)}Bu(\tau)d\tau + Du(t), \quad \bar{g}(t) := Ce^{A t}B$$

$$y_f(t) = y_{zs}(t) = \int_0^t \bar{g}(t-\tau)u(\tau)d\tau + Du(t) = \int_0^t \bar{g}(\tau)u(t-\tau)d\tau + Du(t)$$

Corollary

A SISO system (\star) is BIBO if and only if $\bar{g}(t) = Ce^{A t}B$ is absolutely integrable in $[0, \infty)$ or

$$\int_0^{\infty} |\bar{g}(t)|dt \leq M < \infty$$

$$\begin{cases} \dot{x} = Ax + Bu, \\ y = Cx + Du, \end{cases} \quad x(t_0) = x_0 \in \mathbb{R}^n, \bar{g}(t) := Ce^{A^t}B$$

$$y_f(t) = y_{zs}(t) = \int_0^t \bar{g}(t-\tau)u(\tau)d\tau + Du(t) = \int_0^t \bar{g}(\tau)u(t-\tau)d\tau + Du(t)$$

$$\hat{g}(s) = \mathcal{L}[Ce^{A^t}B] = C(sI - A)^{-1}B$$

$$\begin{aligned} \hat{g}(s) &= \frac{\alpha_1 s^{v-1} + \alpha_2 s^{v-2} + \dots + \alpha_v}{(s - \lambda_1)^{m_1} (s - \lambda_2)^{m_2} \dots (s - \lambda_k)^{m_k}} \\ &= \frac{a_{11}}{s - \lambda_1} + \frac{a_{12}}{(s - \lambda_1)^2} + \dots + \frac{a_{1m_1}}{(s - \lambda_1)^{m_1}} + \dots \\ &\quad + \frac{a_{k1}}{s - \lambda_k} + \frac{a_{k2}}{(s - \lambda_k)^2} + \dots + \frac{a_{km_k}}{(s - \lambda_k)^{m_k}} \end{aligned}$$

$$\begin{aligned} \bar{g}(t) = \mathcal{L}^{-1}[\hat{g}(s)] &= \bar{a}_{11}e^{\lambda_1 t} + \bar{a}_{12}te^{\lambda_1 t} + \dots + \bar{a}_{1m_1}t^{m_1-1}e^{\lambda_1 t} + \dots + \\ &\quad \bar{a}_{k1}e^{\lambda_k t} + \bar{a}_{k2}te^{\lambda_k t} + \dots + \bar{a}_{km_k}t^{m_k-1}e^{\lambda_k t} \end{aligned}$$

BIBO stability of SISO LTI systems

$$\bar{g}(t) = \mathcal{L}^{-1}[\hat{g}(s)] = \bar{a}_{11}e^{\lambda_1 t} + \bar{a}_{12}te^{\lambda_1 t} + \dots + \bar{a}_{1m_1}t^{m_1-1}e^{\lambda_1 t} + \dots + \bar{a}_{k1}e^{\lambda_k t} + \bar{a}_{k2}te^{\lambda_k t} + \dots + \bar{a}_{km_k}t^{m_k-1}e^{\lambda_k t}$$

Notice that some of \bar{a}_{ij} 's can be zero.

Corollary

A SISO system (\star) is BIBO if and only if $\bar{g}(t) = Ce^{\lambda t}B$ is absolutely integrable in $[0, \infty)$ or

$$\int_0^{\infty} |\bar{g}(t)| dt \leq M < \infty$$

- 1 If all the poles p_i of $\hat{g}(s)$ have strictly negative real parts, then $\bar{g}(t) \rightarrow 0$ as $t \rightarrow \infty$. Then the system is BIBO stable.
- 2 If at least one of the $\hat{g}(s)$'s poles has a pole p_i with zero or positive real part, then $|\bar{g}(t)|$ does not converge to zero and the system is **not** BIBO stable.

Theorem (Frequency domain BIBO condition)

The following statements are equivalent

- 1 The LTI system (\star) is uniformly BIBO stable.
- 2 Every pole of the transfer function of the system has strictly negative real parts.

BIBO stability of MIMO LTI systems

Theorem (Time domain BIBO condition)

A MIMO LTI system with impulse response matrix $G(t) = [g_{ij}(t)]$ is BIBO stable, if and only if every $g_{ij}(t)$ is absolutely integrable in $[0, \infty)$.

Theorem (Frequency domain BIBO condition)

A MIMO LTI system with proper rational transfer function $\hat{G}(s) = [\hat{g}_{ij}] = [\hat{g}_{ij}] + D$ is BIBO stable iff every pole of \hat{g}_{ij} has negative real parts.

Corollary

When the LTI system is exponentially stable, then it must be also BIBO stable. *The converse of this statement is not true!*

Recall that, for a system $\{A, B, C, D\}$, the poles of its transfer function are \subseteq of set of eigenvalues of system matrix A

The following two systems are internally unstable but they are BIBO stable

- $\dot{x} = \begin{bmatrix} -2 & 5 \\ 0 & 3 \end{bmatrix} x + \begin{bmatrix} 4 \\ 0 \end{bmatrix} u$, $y = [7 \quad 8] x + 1.5u$: eigenvalues = $\{-2, 3\}$, poles of transfer function = $\{-2\}$
- $\dot{x} = \begin{bmatrix} 1 & 0 \\ 0 & -2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$, $y = [1 \quad 1] x$: eigenvalues = $\{1, -2\}$, poles of transfer function = $\{-2\}$

$$\begin{cases} \dot{x} = A(t)x + B(t)u, \\ y = C(t)x + D(t)u, \end{cases}, \quad (*)$$

$$\bar{G}(t, \tau) := C(\tau)\phi(t, \tau)B(\tau) = [\bar{g}_{ij}(t, \tau)]$$

$$\begin{aligned} y_f(t) = y_{zs}(t) &= \int_0^t C(\tau)\phi(t, \tau)B(\tau)u(\tau)d\tau + D(t)u(t) = \\ &= \int_0^t \bar{G}(t, \tau)u(\tau)d\tau + D(t)u(t) = \int_0^t \bar{G}(\tau)u(t, \tau)d\tau + D(t)u(t) \end{aligned}$$

Theorem (Time domain BIBO stability condition for LTV)

The following statements are equivalent

- 1 The LTV system (*) is uniformly BIBO stable.
- 2 Every entry of $D(\cdot)$ is uniformly bounded and

$$\int_0^t |\bar{g}_{ij}(t, \tau)|d\tau \leq M_{ij} < \infty.$$

- Controllable and reachable subspaces

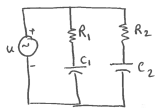
$$\begin{cases} \dot{x} = A(t)x + B(t)u, \\ y = C(t)x + D(t)u, \end{cases} \quad x(t_0) = x_0 \in \mathbb{R}^n$$

- Can we steer the system states from zero initial conditions to any place in the space in finite time? **If not every point, what subset of points we can go to?**
- Can we steer the system states from any arbitrary point in the space to the origin in finite time? **If not every point, what subset of points we can steer to origin?**

Controllable and reachable subspaces: example

$$\dot{x} = \begin{bmatrix} -\frac{1}{R_1 C_1} & 0 \\ 0 & -\frac{1}{R_2 C_2} \end{bmatrix} x + \begin{bmatrix} \frac{1}{R_1 C_1} \\ \frac{1}{R_2 C_2} \end{bmatrix} u$$

$$x(t) = \begin{bmatrix} e^{-\frac{t}{R_1 C_1}} x_1(0) \\ e^{-\frac{t}{R_2 C_2}} x_2(0) \end{bmatrix} + \int_0^t \begin{bmatrix} e^{-\frac{t-\tau}{R_1 C_1}} \frac{1}{R_1 C_1} \\ e^{-\frac{t-\tau}{R_2 C_2}} \frac{1}{R_2 C_2} \end{bmatrix} u(\tau) d\tau$$



x_1 : voltage of C_1
 x_2 : voltage of C_2

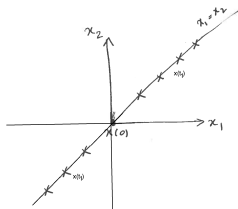
when $R_1 C_1 = R_2 C_2 = 1/\omega$: $x(t) = e^{-\omega t} x(0) + \omega \int_0^t e^{-\omega(t-\tau)} u(\tau) d\tau \begin{bmatrix} 1 \\ 1 \end{bmatrix}$

Starting from $x(0) = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$, the question here is what points in the space are reachable in finite time $t_1 > 0$ by applying a control input.

$$x(t) = e^{-\omega t} \begin{bmatrix} 0 \\ 0 \end{bmatrix} + \omega \int_0^t e^{-\omega(t-\tau)} u(\tau) d\tau \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \underbrace{\omega \int_0^t e^{-\omega(t-\tau)} u(\tau) d\tau}_{\alpha(t)} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

$$x(t_1) = \alpha(t) \begin{bmatrix} 1 \\ 1 \end{bmatrix}, \quad \alpha(t_1) := \omega \int_0^{t_1} e^{-\omega(t-\tau)} u(\tau) d\tau$$

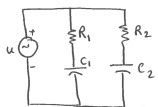
$$\text{reachable points: } x(t_1) = \left\{ \alpha \begin{bmatrix} 1 \\ 1 \end{bmatrix} : \alpha \in \mathbb{R} \right\}, \quad \forall t_1 > t_0 \geq 0.$$



Controllable and reachable subspaces: example

$$\dot{x} = \begin{bmatrix} -\frac{1}{R_1 C_1} & 0 \\ 0 & -\frac{1}{R_2 C_2} \end{bmatrix} x + \begin{bmatrix} \frac{1}{R_1 C_1} \\ \frac{1}{R_2 C_2} \end{bmatrix} u$$

$$x(t) = \begin{bmatrix} e^{-\frac{t}{R_1 C_1}} x_1(0) \\ e^{-\frac{t}{R_2 C_2}} x_2(0) \end{bmatrix} + \int_0^t \begin{bmatrix} e^{-\frac{t-\tau}{R_1 C_1}} \frac{1}{R_1 C_1} \\ e^{-\frac{t-\tau}{R_2 C_2}} \frac{1}{R_2 C_2} \end{bmatrix} u(\tau) d\tau$$



x_1 : voltage of C_1
 x_2 : voltage of C_2

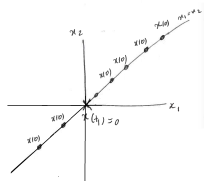
when $R_1 C_1 = R_2 C_2 = 1/\omega$: $x(t) = e^{-\omega t} x(0) + \omega \int_0^t e^{-\omega(t-\tau)} u(\tau) d\tau \begin{bmatrix} 1 \\ 1 \end{bmatrix}$

The question here is what initial conditions $x(0) = x_0 \in \mathbb{R}^2$ can be steered to origin in finite time $t_1 > 0$ ($x(t_1) = 0$) using control inputs. Therefore, our objective is to find x_0 that satisfy

$$0 = e^{-\omega t_1} x(0) + \omega \int_0^{t_1} e^{-\omega(t_1-\tau)} u(\tau) d\tau \begin{bmatrix} 1 \\ 1 \end{bmatrix}.$$

As you can see, our objective is only achievable if $x(0)$ is aligned with $\begin{bmatrix} 1 \\ 1 \end{bmatrix}$, note that

$$x(0) = -\underbrace{\omega e^{\omega t_1} \int_0^{t_1} e^{-\omega(t_1-\tau)} u(\tau) d\tau}_{\beta(t_1)} \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$



Therefore, the 'controllable' initial conditions are $x_0 = \left\{ \beta \begin{bmatrix} 1 \\ 1 \end{bmatrix} : \beta \in \mathbb{R} \right\}$, $\forall t_1 > t_0 \geq 0$.